

A CLASS OF ISI-FREE AND BANDLIMITED PULSES

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Abstract: A novel class of ISI-free pulses is presented and investigated. We propose and investigate a class of new Nyquist pulses produced by Nyquist filters characteristics obtained from combining two types of characteristics with odd-symmetry and a linear characteristic. They show comparable or better ISI performance in the presence of sampling errors, as compared with some recently proposed pulses.

Keywords: intersymbol interference, Nyquist filter, error probability.

1. Introduction

The most common pulse used in telecommunication is the so-called *raised cosine* (RC) pulse. Recently, some new ISI free pulses [2], [3] were proposed. They show better performance than RC with respect to timing error sensitivity.

This paper presents a class of new Nyquist pulses that perform better than RC and FE pulse [2].

2. A class of new Nyquist pulses

We propose a class of new Nyquist pulses with piecewise characteristics, that are illustrated in Figure 1.

The Nyquist filter characteristic is obtained from combining two types of characteristics with odd-symmetry and a linear characteristic. Here $H_i(f)$ and $G_i(f)$ are the family of parabolic and cubic ramps.

For i odd they show odd symmetry around B and their definition is:

$$S_i(f) = \begin{cases} 1, & |f| \leq B(1-a) \\ L(f) & B(1-a) \leq |f| \leq B(1-c) \\ G_i(|f| - B(1-a)) & B(1-c) \leq |f| \leq B(1-b) \\ H_i(f) & B(1-b) \leq |f| \leq B(1+b) \\ 1 - G_i(|f| - B(1-a)) & B(1+b) \leq |f| \leq B(1+c) \\ L(f) & B(1+c) \leq |f| \leq B(1+a) \\ 0, & B(1+a) \leq |f| \end{cases} \quad (1)$$

$$G_i(f) = 1 + \frac{(f)^i}{2B^i a(a-b)^{(i-1)}} \quad (2)$$

$$H_i(f) = 1 + \frac{(B-|f|)^i}{2B^i ab^{(i-1)}} \quad (3)$$

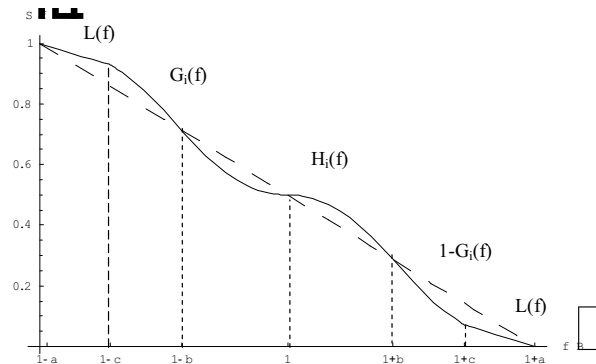


Figure 3. Proposed filter characteristic.

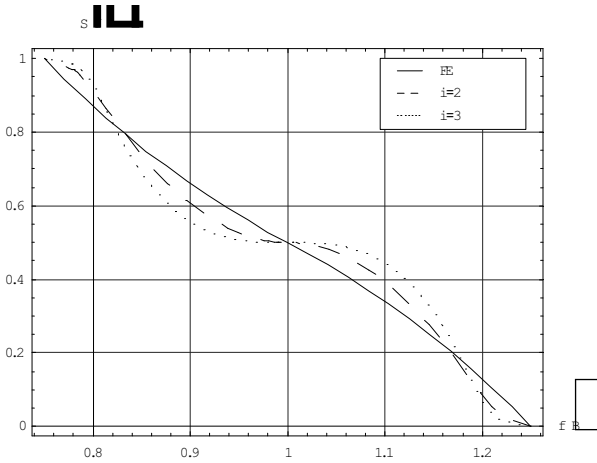


Figure 4. Frequency characteristics for an excess bandwidth $a=0.35$ (positive frequencies).

$$L(f) = \frac{(1+a)(a-c)^{i-1}}{2a(a-b)^{i-1}} - \frac{(b-c)(2a-b-c)^{i-2}}{(a-b)^{i-1}} - \frac{f}{2a(a-b)^{i-1}B} \frac{f}{(a-c)^{i-1}}$$

$$L'(f) = \frac{(1+a)(a-c)^{i-1}}{2a(a-b)^{i-1}} - \frac{f}{2a(a-b)^{i-1}B} \frac{f}{(a-c)^{i-1}} \quad (4)$$

For i even, the vestigial symmetry is obtained by choosing $H_i(f)$ for the frequency interval $B(1-\alpha) \leq f \leq B$ and $1-H_i(f)$ for $B \leq f \leq B(1+\alpha)$. The expressions were derived imposing continuity conditions at $f = B(1-c)$, $f = B(1-b)$, $f = B(1+b)$ and $f = B(1+c)$ and a value of 0.5 at $f = B$.

$L(f)$ and $L'(f)$ are two linear ramps which show characteristics like $x - \frac{f}{y}$, $x \in \mathbb{N}$; $y \in \mathbb{N}$.

This technique is illustrated in Figure 1.

Figure 2 illustrates this types of new Nyquist filter characteristics for $i=2$ and 3 , together with the *flipped exponential* (FE) defined in [2], taken as a reference.

Since they are more concave than the FE pulse, a decrease of the first side lobe in time domain is to be expected, as shown in Fig.3, where a time-scaled replica of pulses is represented for $a = 0,25$. The impulse responses $s_i(t)$ are given in the **Appendix A**.

Their behavior around $t/T = 3,4,\dots$ is more flat, which accounts for their better properties

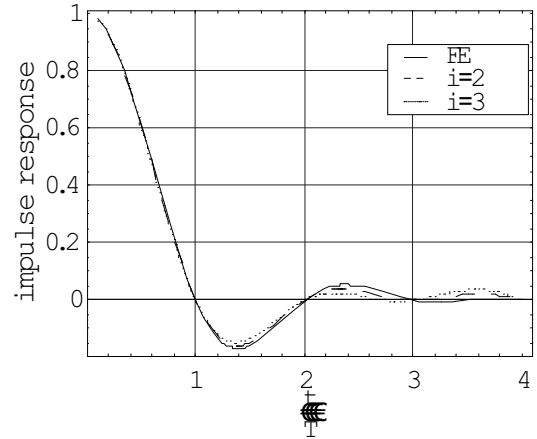


Figure 5. Impulse responses ($a=0.25$).

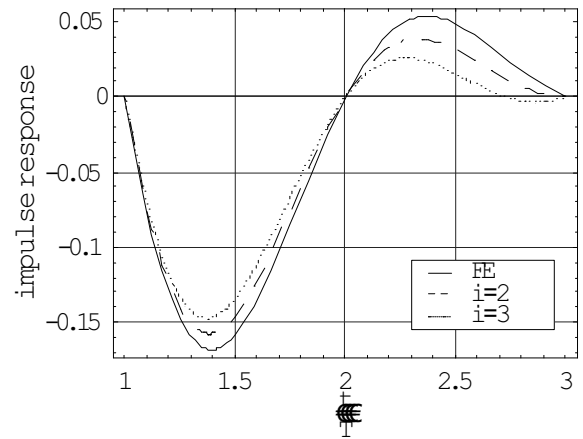


Figure 6. Impulse responses ($a=0.25$).

regarding the error probability when sampled with a small time offset.

A look at Fig.4 that further illustrates the decay of impulse responses reveals that the new pulse defined by (1), (2) and (3) with $i=2$ and 3 , follows closely the FE pulse.

3. Error probability

The error probability is calculated using the method of [13] for these proposed pulses are illustrated in Table I, together with those for FE pulse.

The figures 5 and 6 illustrate the probabilities of error for the proposed pulses and different values of parameters b and c , with different timing offsets.

4. Conclusions

A new class of ISI-free and bandlimited pulses generated by Nyquist filter characteristics resulted from combining two types of characteristics with odd-symmetry and a linear characteristic was presented. The pulses show decreased symbol error probability in the presence of timing error as compared with the

FE pulse [2] with the same roll-off factor a . Its transmission properties were thoroughly investigated and show that the pulses have practical importance.

They achieve best performance if the width of the concave shaped region in the frequency range $(B(1-\alpha), B)$ is maximized at the expense of convex and linear regions.

Table 1. ISI error probability of several Nyquist pulses for $N=2^{10}$ interfering symbols and $\text{SNR} = 15$ dB

P_e	$B=1$			$t/T_B = \pm 0.05$	$t/T_B = \pm 0.1$	$t/T_B = \pm 0.2$
	a	b	c			
FE	0.25			$5.81166 \cdot 10^{-8}$	$1.29804 \cdot 10^{-6}$	$3.56785 \cdot 10^{-4}$
	0.35			$3.92526 \cdot 10^{-8}$	$5.40211 \cdot 10^{-7}$	$1.01287 \cdot 10^{-4}$
	0.5			$2.41342 \cdot 10^{-8}$	$1.85795 \cdot 10^{-7}$	$2.08778 \cdot 10^{-5}$
$s_2(t)$ $i=2$	0.25	0.24	0.241	$5.41937 \cdot 10^{-8}$	$1.13714 \cdot 10^{-6}$	$3.03865 \cdot 10^{-4}$
			0.249	$5.39275 \cdot 10^{-8}$	$1.1306 \cdot 10^{-6}$	$3.02198 \cdot 10^{-4}$
		0.231	0.24	$5.48578 \cdot 10^{-8}$	$1.7396 \cdot 10^{-6}$	$3.17975 \cdot 10^{-4}$
		0.239		$5.43497 \cdot 10^{-8}$	$1.4329 \cdot 10^{-6}$	$3.05961 \cdot 10^{-4}$
	0.35	0.34	0.341	$3.62412 \cdot 10^{-8}$	$4.779 \cdot 10^{-7}$	$9.12275 \cdot 10^{-5}$
			0.349	$3.61147 \cdot 10^{-8}$	$4.75944 \cdot 10^{-7}$	$9.08504 \cdot 10^{-5}$
		0.331	0.34	$3.66311 \cdot 10^{-8}$	$4.90692 \cdot 10^{-7}$	$9.48001 \cdot 10^{-5}$
		0.339		$3.63241 \cdot 10^{-8}$	$4.79941 \cdot 10^{-7}$	$9.17449 \cdot 10^{-5}$
	0.5	0.49	0.491	$2.23906 \cdot 10^{-8}$	$1.71805 \cdot 10^{-7}$	$2.3126 \cdot 10^{-5}$
			0.499	$2.23363 \cdot 10^{-8}$	$1.71282 \cdot 10^{-7}$	$2.30509 \cdot 10^{-5}$
		0.481	0.49	$2.25801 \cdot 10^{-8}$	$1.75637 \cdot 10^{-7}$	$2.37756 \cdot 10^{-5}$
		0.489		$2.24289 \cdot 10^{-8}$	$1.72398 \cdot 10^{-7}$	$2.32228 \cdot 10^{-5}$
$s_3(t)$ $i=3$	0.25	0.24	0.241	$5.17281 \cdot 10^{-8}$	$1.06603 \cdot 10^{-6}$	$2.87411 \cdot 10^{-4}$
			0.249	$5.17726 \cdot 10^{-8}$	$1.06826 \cdot 10^{-6}$	$2.88272 \cdot 10^{-4}$
		0.231	0.24	$5.31151 \cdot 10^{-8}$	$1.2737 \cdot 10^{-6}$	$3.09256 \cdot 10^{-4}$
		0.239		$5.18662 \cdot 10^{-8}$	$1.07197 \cdot 10^{-6}$	$2.89491 \cdot 10^{-4}$
	0.35	0.34	0.341	$3.48359 \cdot 10^{-8}$	$4.70575 \cdot 10^{-7}$	$9.7711 \cdot 10^{-5}$
			0.349	$3.48554 \cdot 10^{-8}$	$4.71236 \cdot 10^{-7}$	$9.79258 \cdot 10^{-5}$
		0.331	0.34	$3.55853 \cdot 10^{-8}$	$4.91119 \cdot 10^{-7}$	$1.03271 \cdot 10^{-4}$
		0.339		$3.49121 \cdot 10^{-8}$	$4.72606 \cdot 10^{-7}$	$9.82458 \cdot 10^{-5}$
	0.5	0.49	0.491	$2.20655 \cdot 10^{-8}$	$1.79094 \cdot 10^{-7}$	$3.06125 \cdot 10^{-5}$
			0.499	$2.20732 \cdot 10^{-8}$	$1.79267 \cdot 10^{-7}$	$3.06597 \cdot 10^{-5}$
		0.481	0.49	$2.24053 \cdot 10^{-8}$	$1.85768 \cdot 10^{-7}$	$3.18409 \cdot 10^{-5}$
		0.489		$2.21005 \cdot 10^{-8}$	$1.79772 \cdot 10^{-7}$	$3.07321 \cdot 10^{-5}$

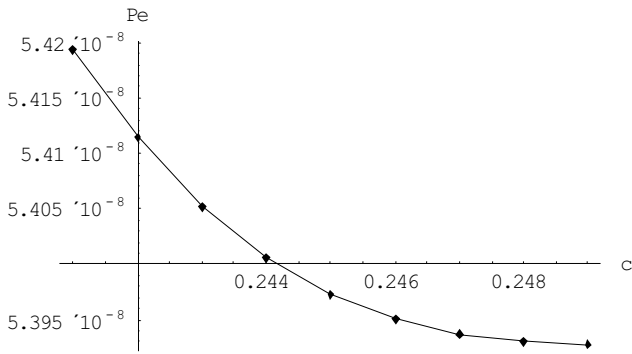


Figure 5a.

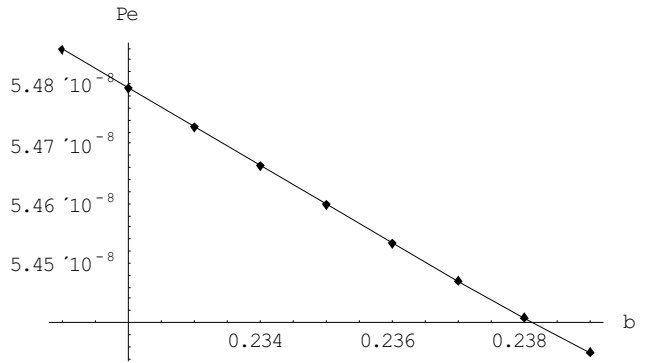


Figure 5b.

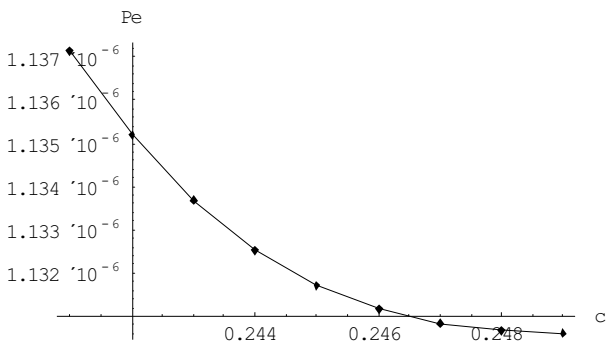


Figure 5c.

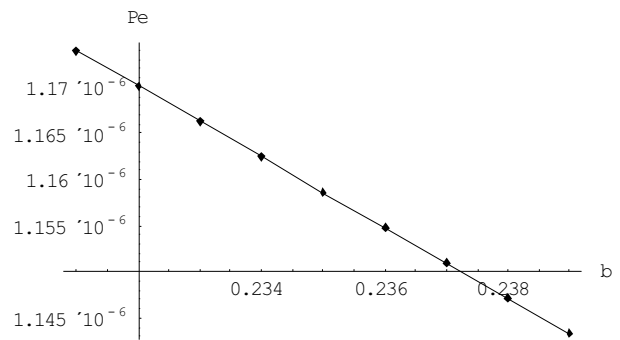


Figure 5d.

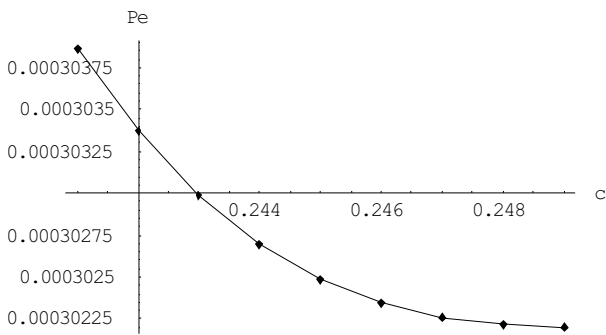


Figure 5e.

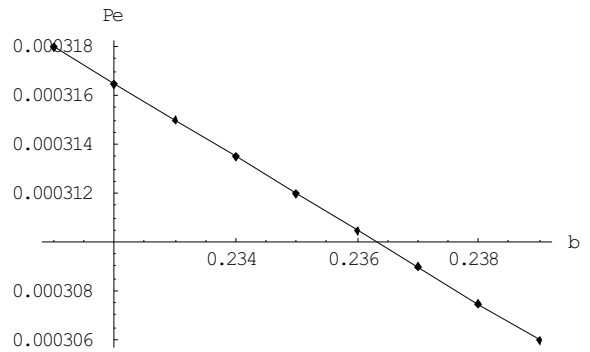


Figure 5f.

Figure 7. Error probability for $s_2(t)$; 5a) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.05$
 5b) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.05$;
 5c) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.1$;
 5d) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.1$;
 5e) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.2$;
 5f) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.2$;

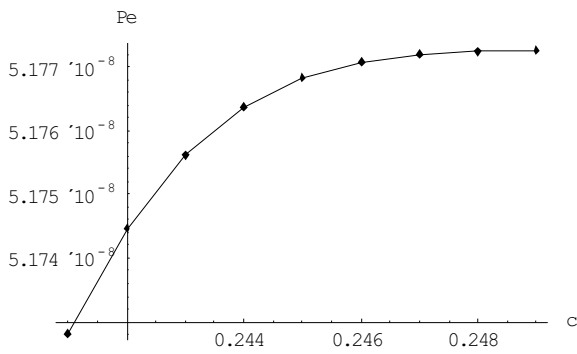


Figure 6a.

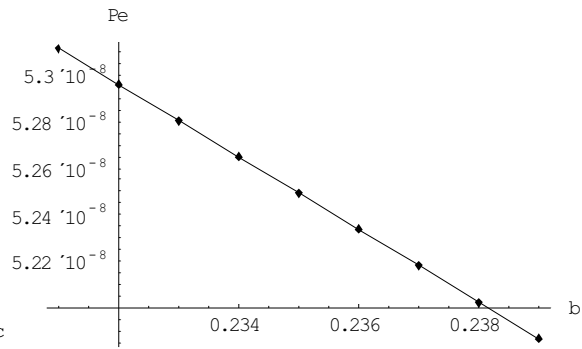


Figure 6b.

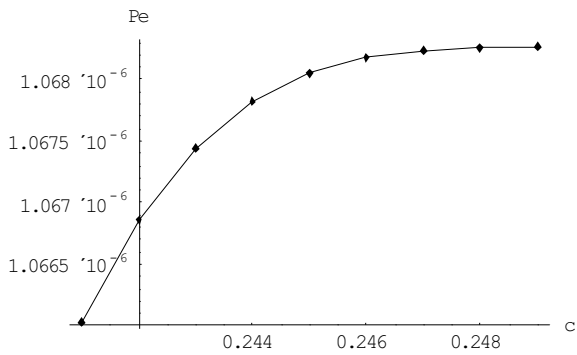


Figure 6c.

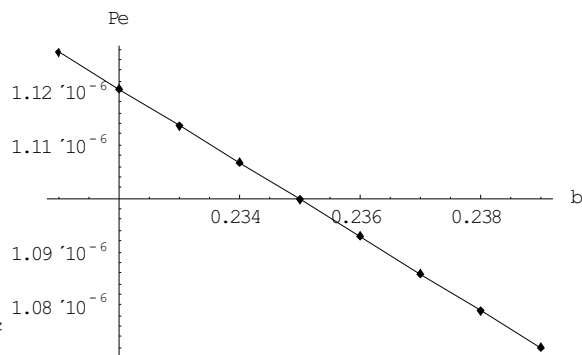


Figure 6d.

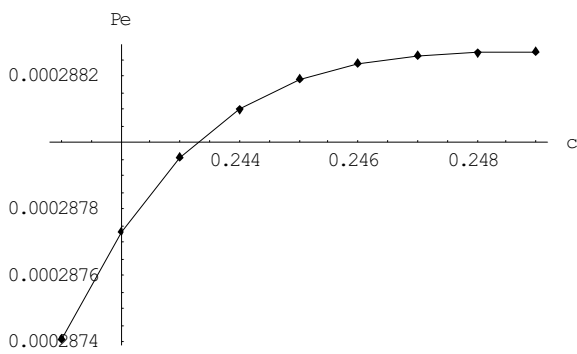


Figure 6e.

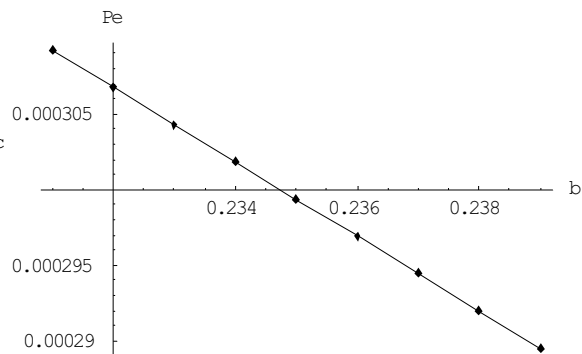


Figure 6f.

Figure 8. Error probability for $s_3(t)$;

- 6a) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.05$
- 6b) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.05$;
- 6c) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.1$;
- 6d) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.1$;
- 6e) $a=0.25, b=0.24, c \in [0.241, 0.249]$; $t/T_B=0.2$;
- 6f) $a=0.25, b \in [0.231, 0.239], c=0.24$; $t/T_B=0.2$;

5. References

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Appendix A

$$s_2(t) = \frac{1}{a(a-b)bB^2\pi^3t^3} \left(\frac{1}{2}bB(a-c)\pi \cos((-1+a)B\pi) + (a-b)bB\pi \cos((1+b)B\pi) + \frac{1}{2}abB\pi \cos((-1+c)B\pi) - \frac{1}{2}bBc\pi \cos((-1+c)B\pi) + 2bB(-a+c)\pi \cos(B(1+c)\pi) - 2a \sin(B\pi) + 2b \sin(B\pi) - a \sin((-1+b)B\pi) + a \sin((1+b)B\pi) + b \sin((1+b)B\pi) - \frac{1}{2}a^2bB^2\pi^2t^2 \sin((1+b)B\pi) + ab^2B^2\pi^2t^2 \sin((1+b)B\pi) - \frac{1}{2}b^3B^2\pi^2t^2 \sin((1+b)B\pi) + b \sin((-1+c)B\pi) + 2b \left(-1 + \frac{1}{2}B^2(a-c)^2\pi^2t^2 \right) \sin((1+c)B\pi) \right)$$

$$s_3(t) = \frac{1}{a(a-b)^2b^2B^3\pi^4t^4} \left(\sin(B\pi) \left(3a(a-b)bB\pi \cos(bB\pi) + 3b^2B(-a+c)\pi \cos(bB\pi) + \frac{1}{2}a^2b^2B^2\pi^2t^2 \sin(aB\pi) - ab^2B^2c\pi^2t^2 \sin(aB\pi) + \frac{1}{2}b^2B^2c^2\pi^2t^2 \sin(aB\pi) - 3a^2 \sin(bB\pi) + 6ab \sin(bB\pi) + 6ab \sin(bB\pi) + b^2 \left(-3 + B^2(a-c)^2\pi^2t^2 \right) \sin(Bc\pi) \right) \right)$$